

Printed Planar Double Inverted-F Antenna With large Frequency Reconfigurability Range

Jean Marie FLOC'H And Imen BEN TRAD

IETR, INSA, 20 avenue des Buttes de Coësmes, 35708 Rennes, France.

ABSTRACT: The use of frequency reconfigurable antennas is extremely important for Cognitive Radio (CR) systems. In this paper, we propose a study and design a multiband planar double inverted F-antenna (PIFA) with frequency agility characteristic. The proposed antenna is suitable for several applications. In fact, a varactor diode has been integrated between the radiating element and the added floating ground plane. By applying a reverse voltage V on the varactor diode and we obtain more than three resonance frequencies with wide tuning ranges. Frequencies can be shifted from 700 to 1450 MHz and 1975 to 2800 MHz. The passive and active antennas have been manufactured and characterized. Simulations and measurements are presented and discussed. The novelty of this antenna is the use of a floating ground plane to realize the reconfigurability.

KEYWORDS: Frequency agility; reconfigurable antenna; Varactor diode; PIFA; double PIFA; large tuning frequency range.

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I. INTRODUCTION

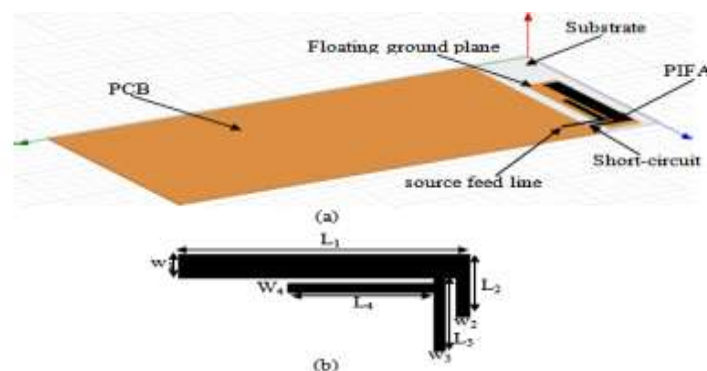
Wireless communication systems must be able to meet increasing requirements of multiple standards, higher data rates and better use of Frequency Spectrum. To overcome these challenges, antennas need to be flexible and adapt to environment changes. Frequency Reconfigurable Antennas are addressing these complex and over changing needs and are consequently becoming a hot topic triggering tremendous research interest [1-6]. This document is an extension of paper [13] listed in References at the end of this document.

The Planar inverted F-antenna (PIFA) has been extensively studied and is widely used in mobile handsets and connected objects. It is simple and cheap to manufacture and has a small footprint. Adding frequency agility property to PIFA antenna provides promising solutions to increasingly complex requirements [7-10].

We have investigated in [11] a PIFA structure with a technique for frequency agility behavior. This technique is based on a concept developed in [12]. In this paper, we have enlarged the frequency reconfigurability ranges by using a double PIFA configuration. The new antenna design is presented in Fig. 1. The purpose of the floating ground plane is to reduce the antenna size and increase the number of resonance frequencies. This antenna can be deployed in several wireless communication applications.

II. PASSIVE ANTENNA

A. Antenna design



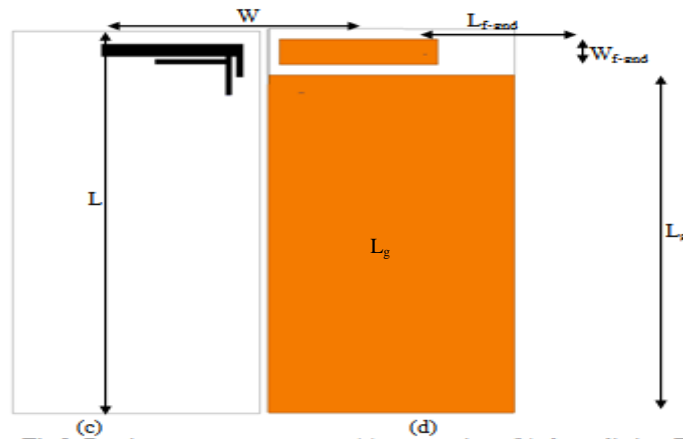


Fig.1. Passive antenna geometry: (a) perspective; (b) the radiating PIFA (c) front side and (d) back side views.

The proposed antenna is printed on FR4 substrate with a thickness $h=0.8\text{mm}$ and a relative permittivity $\epsilon_r = 4.4$. The PIFA structure is printed on the top side and short-circuited to the grounded PCB on the backside as presented in Fig.1-a/b. The floating ground plane is added on the backside below the PIFA as shown in Fig.1-c. It enables changing the antenna effective permittivity and increases the electrical length of the radiating section. Thus, the overall size of the antenna is well reduced as explained in [11] and [12]. The simulated results of antenna with (blue) and without (red) floating grounded plane are presented in Fig. 2. It illustrates very well this behavior; we generate lower frequency resonance (1250 MHz to 1150 MHz) and new frequency resonances in the same frequency band. The optimized design parameters are given in Table I.

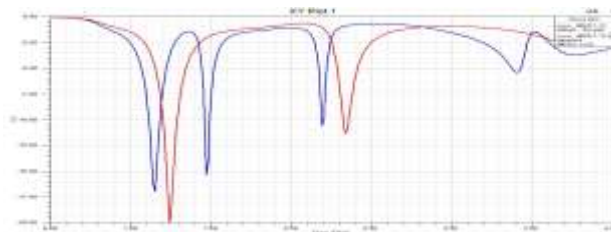


Fig.2. Simulated reflection coefficient S_{11} of the passive antenna with (blue) and without (red) floating grounded plane.

TABLE I. DESIGN PARAMETERS OF THE PASSIVE ANTENNA

Parameter	Value (mm)	Parameter	Value (mm)
L	150	w ₁	5
W	70	w ₂	2
L ₁	40	w ₃	1.6
L ₂	13	L _g	132
L ₃	15	W _{f-gnd}	10
L ₄	20	L _{f-gnd}	45

B. Passive PIFA performances

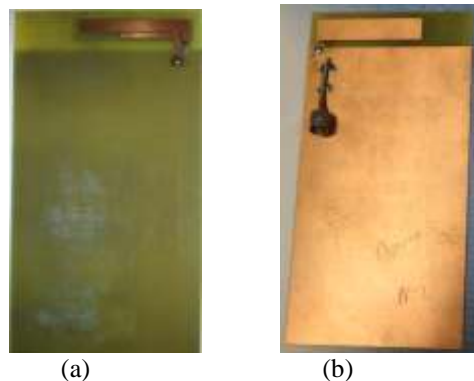


Fig.3. Photos of the prototyped passive antenna: (a) front side and (b) back side views.

The antenna prototype is presented in Fig.3. The passive PIFA has been numerically studied and optimized using the ANSYS-ANSOFT High Frequency Structure Simulator, HFSS-V15. Practically, good agreement is obtained between simulation results in red and measurements in blue as shown in Fig 4. The slight shift observed is the result of manufacturing inaccuracies (the real permittivity of FR4 per example) and we don't take into account the coaxial feeding in the simulation. The measured resonance frequencies are $F_1=1067$, $F_2=1602$ and $F_3=2278$ MHz.

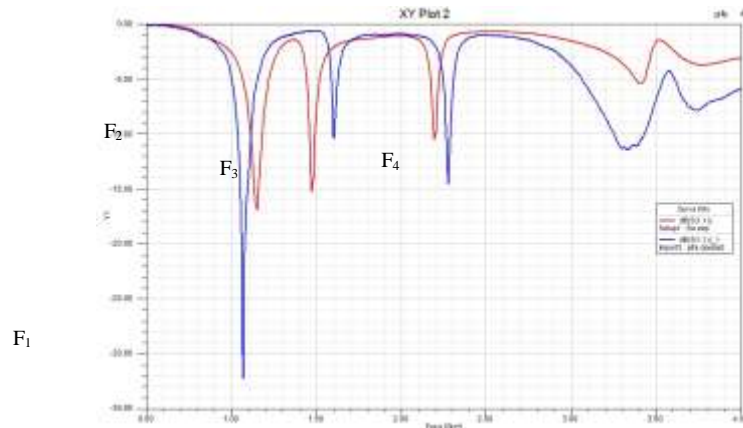
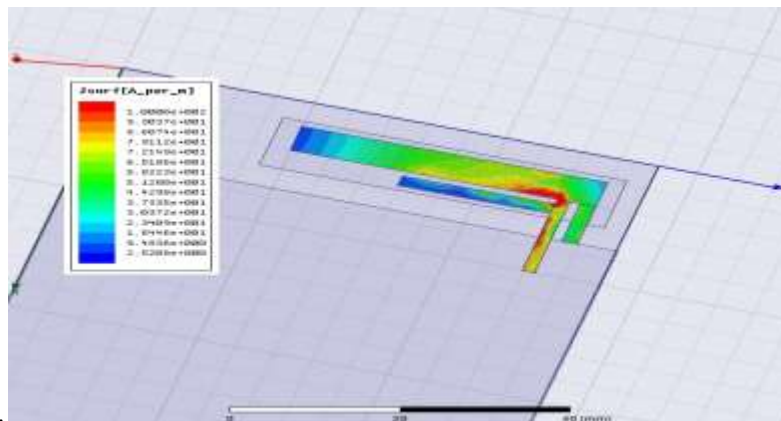
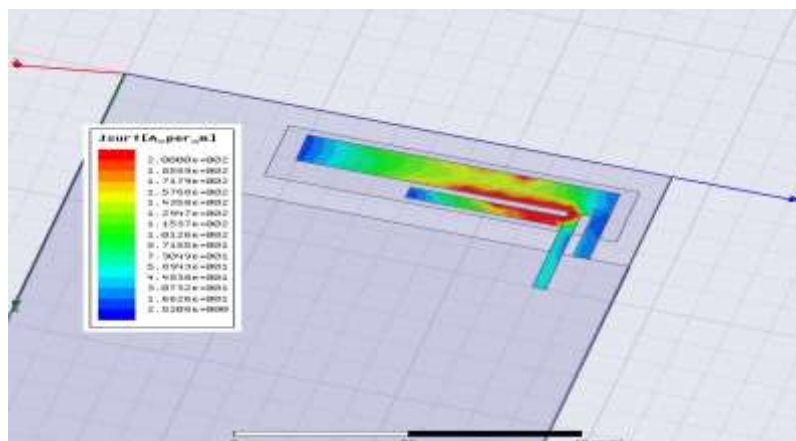


Fig.4. Measured (blue) and simulated (red) reflection coefficient S_{11} of the passive antenna.

Fig. 5 shows the simulated surface current for the three resonant frequencies $F1 = 1150$ MHz, $F2 = 1470$ MHz and $F3 = 2190$ MHz. As expected, we observe a large current on the PIFA bigger arm for $F1 = 1150$ MHz, a large current on both arms for $F2 = 1470$ MHz and a large current on the little arm for $F3 = 2190$ MHz



a) 1150 MHz



b) 1470 MHz

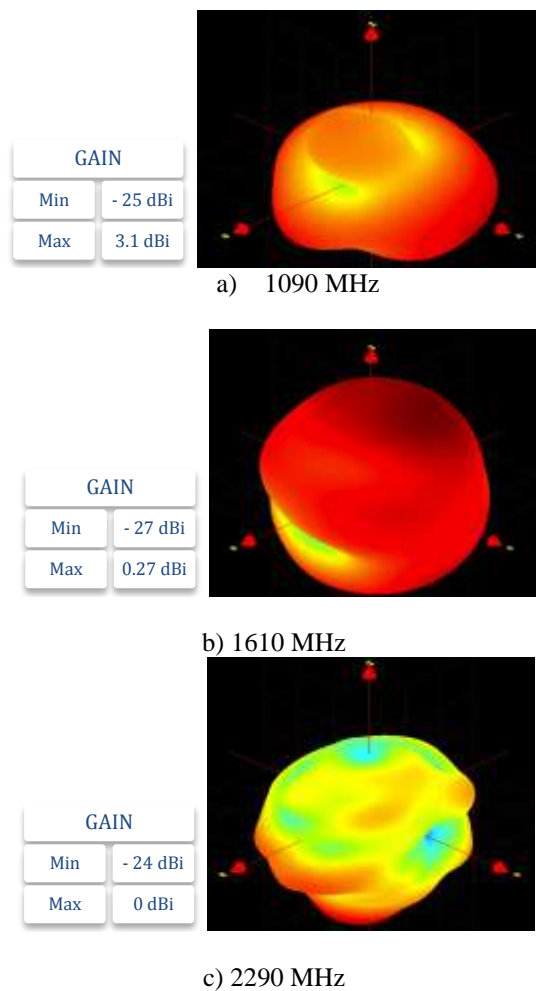


Fig.7. Measured radiation patterns of the passive antenna at the 3 resonant frequencies.

III. FREQUENCY RECONFIGURABLE ANTENNA

The varactor diode used in the active antenna is a BB833 device from Infineon. It is integrated in the corner of the bigger line of the PIFA, to connect it later to the floating ground plane as can be seen on Fig.5. The agility is obtained by applying a variable reverse voltage V to the varactor diode. A $1\text{ K}\Omega$ resistor is used to polarize the diode. The measured reflection coefficient S_{11} for different cases (voltage between 0 to 17 V) are shown in Fig.9.

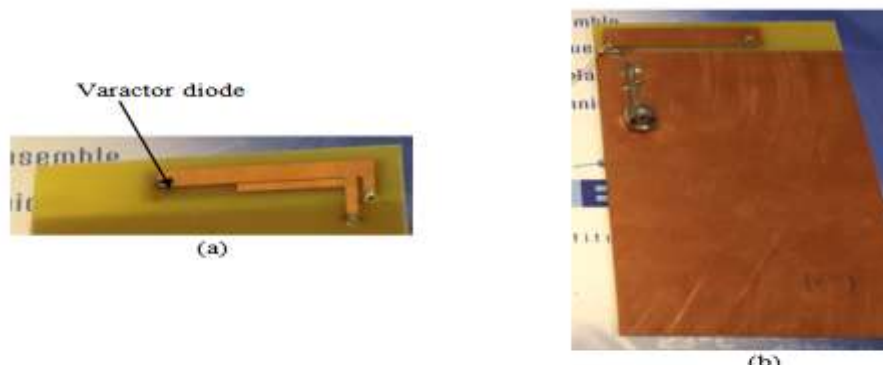


Fig.8. Photos of the fabricated frequency reconfigurable double PIFA: (a) front side and (b) back side views.

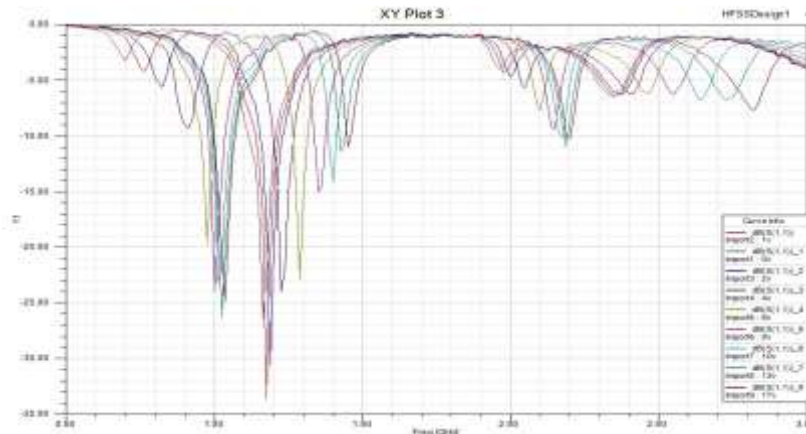


Fig.9. Measured reflection coefficient S_{11} of the frequency reconfigurable antenna.

The addition of avaractor diode to PIFA allows an electronic control of the four resonance frequencies that shift to higher values when increasing the applied reverse bias voltage via a decrease of the capacitor values. We will focus on the first three frequencies F_1 , F_2 and F_3 . The resonance frequency F_1 can be easily moved from 700 to 1037 MHz. Noteworthy is that the input impedance matching was greatly improved while decreasing the capacitor values. For the second frequency F_2 , we obtain a frequency range between 1162 to 1450 MHz. A tuning frequency range of 1975 to 2200 MHz is achieved for the Third resonance frequency F_3 . Measured resonant frequencies are summarized in Table II. Fig.10 presents the measured and simulated reflection coefficient S_{11} for reverse bias voltage $V=13V$ ($C=1pF$). A good agreement can be observed between simulated and measured results.

TABLE II. MEASURED RESONANT FREQUENCIES

Voltage (V)	Capacitor value (pF)	Frequency F_1 (MHz)	Frequency F_2 (MHz)	Frequency F_3 (MHz)
without varactor	-	1067	1602	2278
0	≈20	700	1162	1975
1	9.8	762	1175	1987
2	7	825	1187	2000
4	4	908	1220	2050
6	2.8	975	1287	2100
8	1.8	1000	1350	2150
10	1.4	1025	1400	2175
13	1	1030	1425	2185
17	0.9	1037	1450	2200

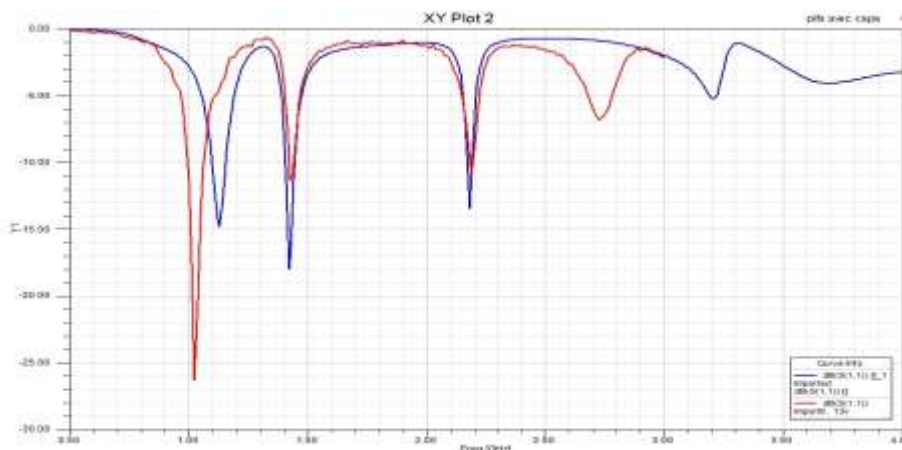


Fig.10. Measured and simulated reflection coefficient S_{11} for $V = 13V$ ($C=1pF$).

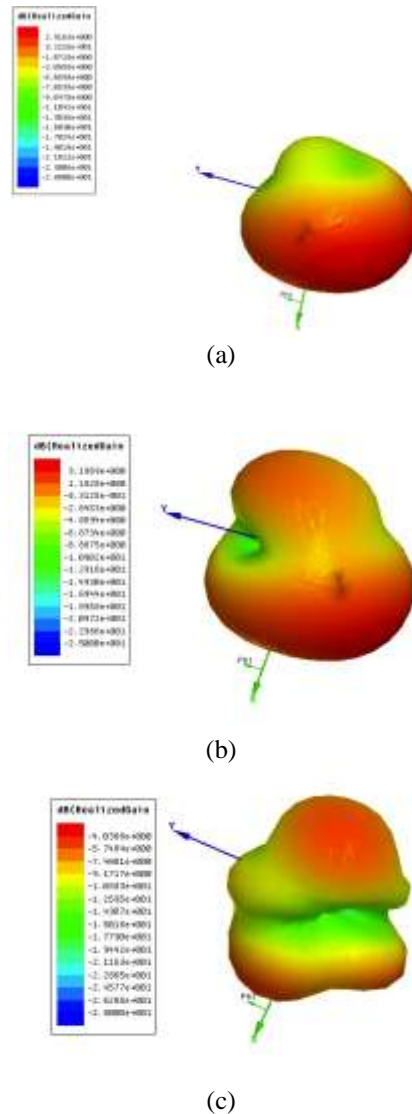
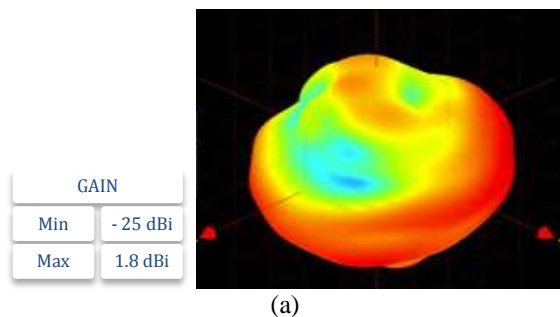


Fig.11. Simulated 3 D radiation patterns for $V = 4V$ ($C=4pF$) at: (a) 1050 , (b) 1300 and (c) 2050 MHz.

The simulated and measured 3D radiation patterns for $V=4V$ ($C=4pF$) are shown in Fig. 11 and Fig. 12, respectively. It can be concluded that the frequency reconfigurable double PIFA presents stable radiation properties with maximum simulated gain of 2.9 dBi at 1050 MHz, 3.2dBi at 1300 MHz and -4 dBi at 2050 MHz. Actual measured gains are $G_{max} = -1.8$ dBi at 908 MHz, 0.6 dBi at 1220 MHz and -9.8 dBi at 2050 MHz. The differences between simulated and measured results can be explained by the fact that we only consider the varactor diode like a capacitance without taking into account the diode losses. Besides, the maximum frequency for the diode is 2 GHz.



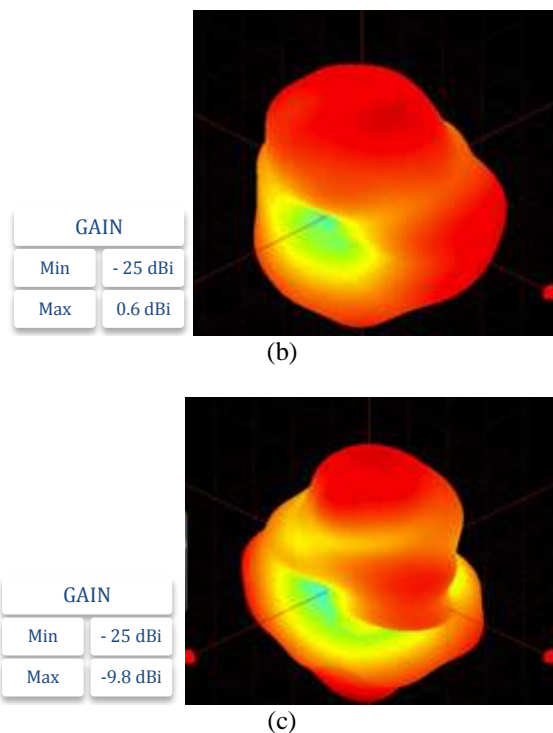


Fig.12. Measured 3 D radiation patterns at: (a) 908; (b) 1220 and (c) 2050 MHz for $V = 4V$ ($C = 4pF$).

We present the measured radiation patterns for more two voltage 0v and 10 Vin the fig. 13 and 14.

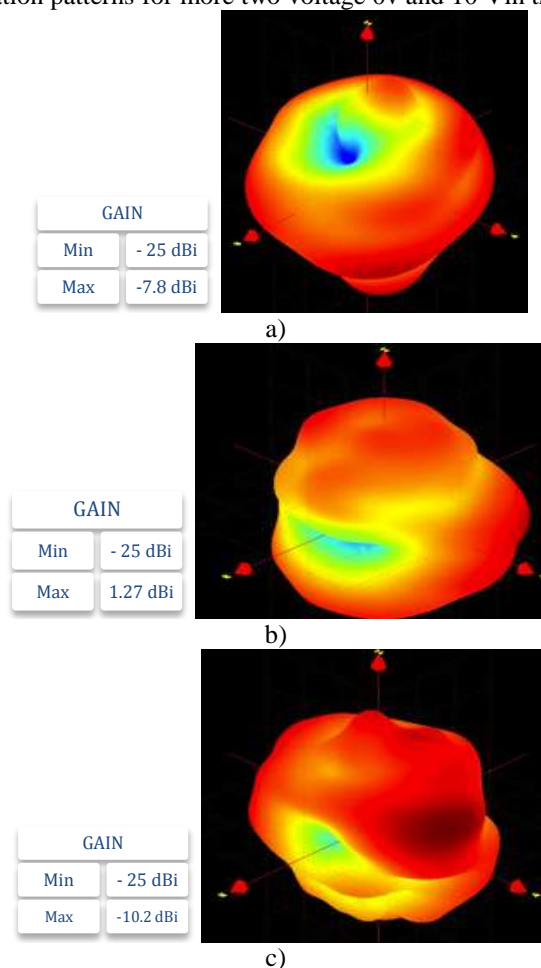


Fig.13. Measured 3 D radiation patterns at: (a) 678; (b) 1160and (c) 1968 MHz for $V = 0V$ ($C = 20pF$).

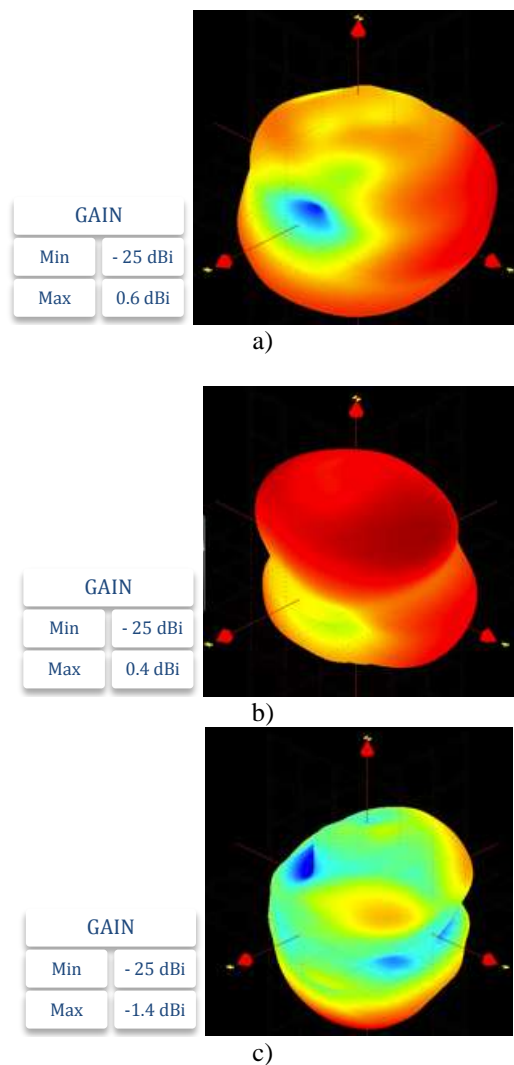


Fig.14. Measured 3 D radiation patterns at: (a) 1020; (b) 1440 and (c) 2110 MHz for $V = 10V$ ($C = 1.4pF$). The antenna efficiency at different voltages and resonance frequencies is shown on this next table:

TABLE III. MEASURED EFFICIENCY

Voltage (V)	Capacitor value (pF)	F ₁	F ₂	F ₃
Without varactor	- -	66%	45%	21%
0	≈20	7%	38%	4%
4	4	19%	37%	4%
10	1.4	30%	27%	10%

We note a degradation of the efficiencies when the capacitance value increases and a better efficiency for the second resonance frequency.

IV. CONCLUSION

A multiband frequency reconfigurable double PIFA has been successfully designed and numerically studied. A varactor diode has been inserted between the double PIFA and the floating ground plane in order to produce the agility property. Four resonance frequencies are electronically controlled and reconfigurable behavior is well obtained between 700 to 1450 MHz and 1975 to 2800 MHz. Measurement results agree well with simulations. The proposed antenna can be used for a large range of applications. This study shows the potential for reconfigurability using a floating ground plane. We are now studying other floating grounded plan structures in order to improve performance in efficiency.

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